

Experimental and Theoretical Study of Parasitic Leakage/Resonance in a K/Ka-Band MMIC Package

Jong-Gwan Yook, *Student Member, IEEE*, Linda P. B. Katehi, *Fellow, IEEE*,
Ranee N. Simons, *Senior Member, IEEE*, and Kurt A. Shalkhauser

Abstract—In this paper, electromagnetic (EM) leakage and spurious resonances in a K/Ka-Band (18–40 GHz) MMIC hermetic package designed for a phase shifter chip are studied using the finite element method (FEM) and the numerical simulation results are compared with the measured data. Both the measured and calculated data indicate several spurious resonances in the 18–24 GHz region and the origin of this phenomenon is identified by virtue of the modeling capability of the FEM. Moreover, the effect of dc bias lines, bond wires, shielding, and the asymmetry of the package on electrical performance are closely examined. In addition, the effect of adding a resistive coating to the inside surface of the package lid and also the use of dielectric packaging materials with very high loss tangent are studied in view of the suppression of the spurious resonances. Finally, design guidelines for the improved package are presented.

I. INTRODUCTION

HIGH performance packages, especially for microwave and millimeter-wave integrated circuit (MMIC) application, should satisfy stringent mechanical, electrical, and environmental requirements. From a mechanical and environmental point of view, a package should provide protection to the internal circuits from the surroundings. Furthermore, packages are required to exhibit minimum insertion loss, good isolation between the ports as well as electromagnetic (EM) shielding for minimum interference (EMI) [1], [2]. Another important electrical requirement of a package is noninvasiveness with respect to circuit performance [3]; a package fails electrically if parasitic cavity resonances substantially deteriorate circuit performance. As a result of all of these requirements, successful development of a high frequency package requires careful design strategies. Recently, low cost high performance MMIC packages have been developed by using approximate equivalent circuit models or experimental data [4], [5]. However, due to the limited accuracy of the modeling tools, the designed packages exhibit a serious performance degradation at higher frequencies. To overcome the above difficulties in designing high frequency/high performance package, frequency and time

domain full-wave electromagnetic tools are applied [6]–[9] for various applications.

The goal of the paper is to study the electrical performance of a K/Ka-band hermetic package designed for a MMIC phase shifter and comprehend the parasitic effects introduced by the package geometry. For thorough understanding of the package performances, the effects of the various features of the package, such as filled metal vias, dc bias lines, bond wires, structural symmetry/asymmetries, and even the effect of the test fixture on the circuit performance are extensively investigated. Furthermore, the use of dielectric packaging materials with high loss tangent and the influence of coating the inside surface of the package lid with a resistive material are also examined. For the EM characterization of the K/Ka-Band MMIC package, a parallelized three-dimensional (3-D) finite element method (FEM), which is optimized on the distributed memory machine (IBM SP2) with task parallelization strategy, is applied [10], [11]. The parallelized 3-D FEM code exhibits near linearly scalable performance improvement due to the frequency independent nature of the frequency domain FEM. To the best of our knowledge, this is the first comprehensive high frequency full-wave treatment for an existing millimeter-wave package.

The modeling effort described herein is divided into three parts. In the first part, the input and the output microstrip lines are symmetrically located with respect to the two planes of symmetry of the package. In addition, the package is considered to be free standing, i.e., totally isolated. Under these assumptions only one quarter of the package needs to be considered and as a result the modeling effort is computationally simpler. In the second case, the input and the output microstrip lines are asymmetrically located with respect to one of the plane of symmetry. This is of more general interest since in real applications the transmission lines on the MMIC chip need not be symmetrically located. In the last part, the microstrip lines are asymmetrically located and the package is placed inside a test fixture. By placing the package inside the test fixture, the interactions between the fields leaking through the package dielectric walls and the surroundings can be modeled. This model is of great practical interest since it points toward ways and means to improve future package performance. The modeled characteristics are compared with experimental results and show very good agreement. In addition, the effects of the three different types of packaging features, such as additional metal-filled vias, resistive coating, and lossy packaging materials, are carefully

Manuscript received March 27, 1996. This work was supported in part by NASA Lewis Research Center under Grant NAG-1807, NSF Grant CDA-92-14296, and the Ford Motor Company.

J.-G. Yook and L. P. B. Katehi are with the Radiation Laboratory, Department of Electrical Engineering and Computer Science, The University of Michigan, Ann Arbor, MI 48109-2122 USA.

R. N. Simons is with the NASA Lewis Research Center, NYMA Group, Cleveland, OH 44135 USA.

K. A. Shalkhauser is with the NASA Lewis Research Center, Cleveland, OH 44135 USA.

Publisher Item Identifier S 0018-9480(96)08544-4.

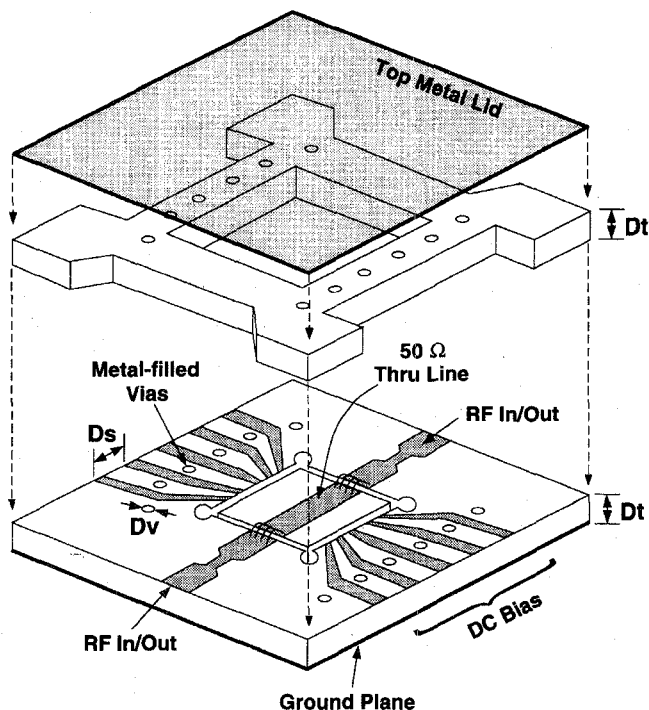


Fig. 1. Schematic diagram of the K/Ka-band hermetic package designed and manufactured for a MMIC phase shifter chip. The diameter of the vertical metal filled vias is $D_v = 0.203$ mm and the distance between these vias is $D_s = 1.016$ mm. The upper and lower alumina layers are each 0.381 mm thick (D_t).

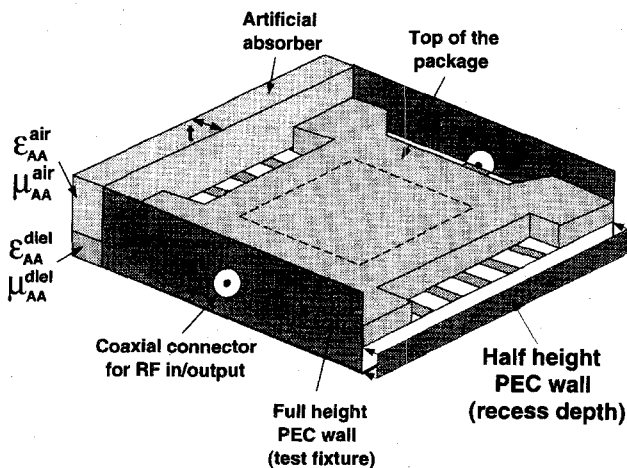


Fig. 2. Schematic showing the location and extent of PEC's in addition to the artificial absorbers which simulate the test fixture and open environment. Two different types of absorbers ($\epsilon_{AA}^{air}, \mu_{AA}^{air}$ and $\epsilon_{AA}^{diel}, \mu_{AA}^{diel}$) are designed and placed in either side of the package facing the dc bias lines (not fully shown in the figure for simplicity).

examined to improve the package performance. This modeling procedure has the potential to predict the performance of other types of packages such as those used in wireless communications which include multichip modules.

II. PACKAGE DESCRIPTION AND MODELING

A K/Ka-band MMIC package fabricated by Hughes Aircraft Company for the NASA Lewis Research Center is shown in Fig. 1. The package has 50 Ω microstrip input/output feed

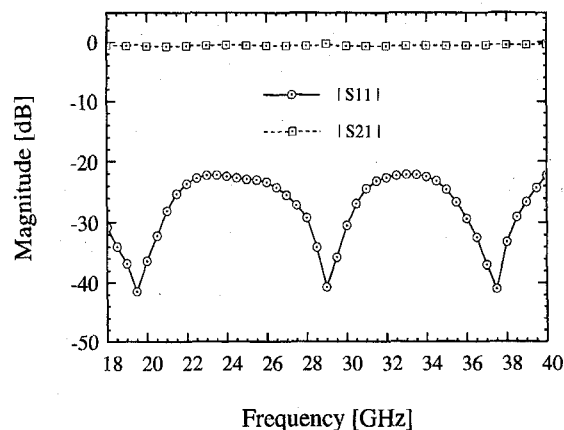


Fig. 3. Computed S -parameters for the isolated symmetric package.

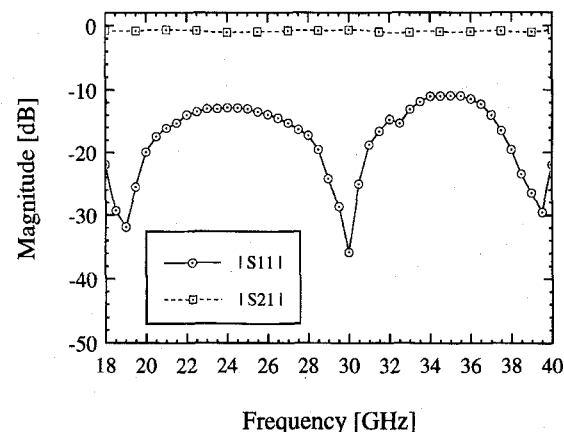


Fig. 4. Computed S -parameters for the isolated asymmetric package.

lines for the RF signal and two sets of five metal lines on either sides for dc bias and control. In addition, a set of twelve metal filled vias tie the top and bottom perfect electric conductor (PEC) ground planes to provide mechanical strength and also serve as an EM shield. The package is fabricated from alumina (92% pure, $\epsilon_r = 9.5$) using the HTCC process. To characterize the package, a 50 Ω through line is placed in the recess as shown in Fig. 1 and is wire bonded to the input/output microstrip feed lines. The peripheral dimensions of the package are $7.112 \times 7.112 \times 1.27$ mm. One can observe from the figure that the metal filled vias and the input/output microstrip feed lines are displaced toward one side of the package introducing a package asymmetry. This asymmetry is attributed to the specific geometry of the MMIC phase shifter which the package intends to house.

While this package provides hermeticity, it does not shield electromagnetically and as a result the packaged circuits are exposed to a semi-open environment. For the simulation of this environment, artificial absorbing layers have been designed using lossy isotropic dielectrics as shown in Fig. 2. The performance of the absorber is controlled by assigning a certain amount of losses in the absorbing material and by specifying its thickness. In this study, we designed two different types of isotropic absorbing materials for air and dielectric side

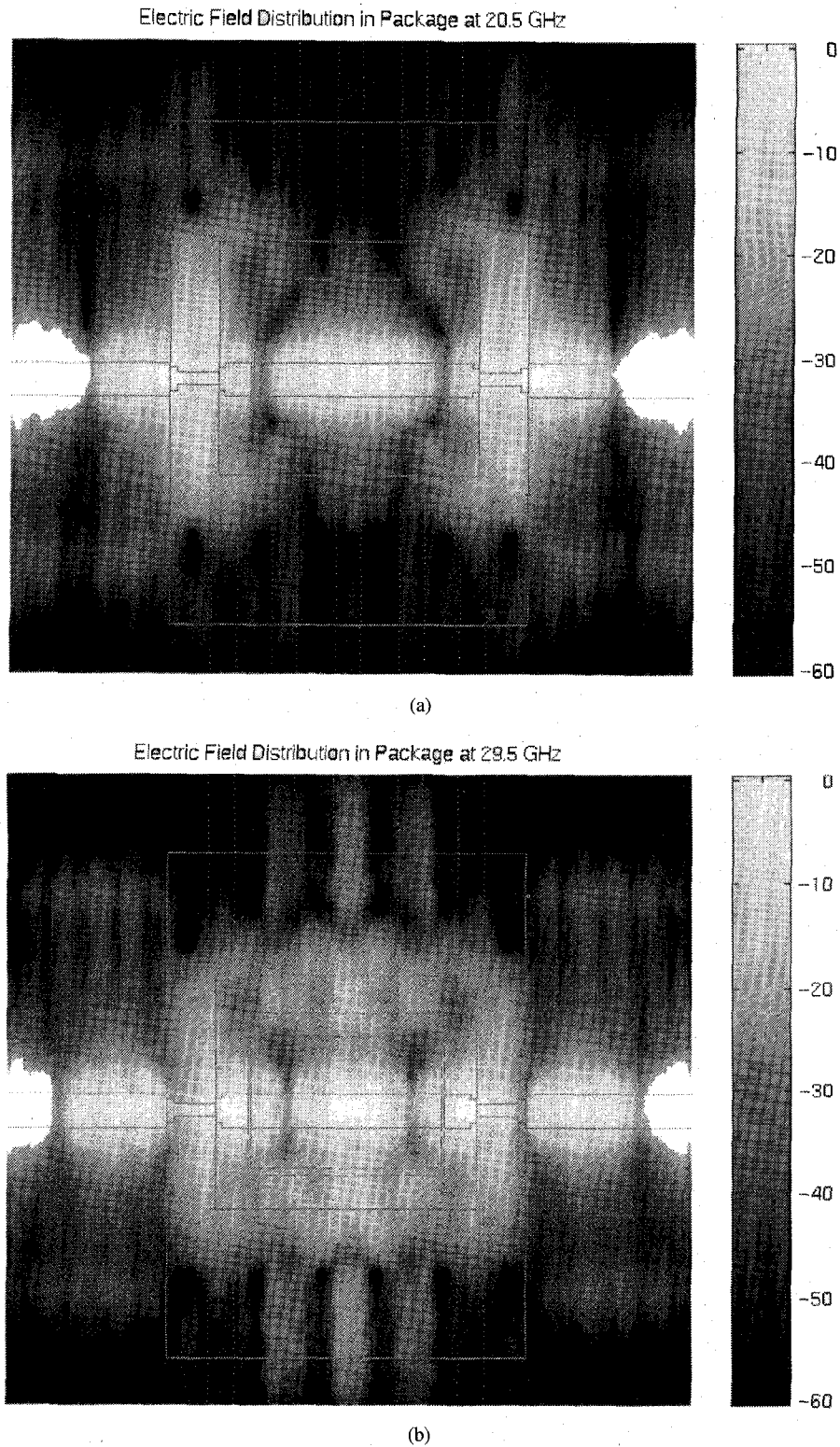


Fig. 5. Computed vertical electric field distribution (dB scale) at (a) $f = 20.5$ GHz and (b) $f = 29.5$ GHz in the isolated asymmetric hermetic package.

terminations. The material parameters are chosen to be $\epsilon_{AA}^{\text{air}} = \mu_{AA}^{\text{air}} = 1.0 + j10.0$ for the air side and $\epsilon_{AA}^{\text{diel}} = 9.5 + j15.0$ and $\mu_{AA}^{\text{diel}} = 1.0 + j1.578$ for the dielectric side. The thickness (t) of the artificial absorbers is assigned to be 0.70 mm to allow enough damping of the fields inside of the absorbers and to minimize the computational domain. It is well known that this type of absorber performs well for near-normal incident fields.

III. NUMERICAL AND EXPERIMENTAL RESULTS AND DISCUSSIONS

A. Modeling of Isolated Symmetric Package

In this case, an isolated symmetric package is designed and modeled. The structural details are very close to that of Fig. 1

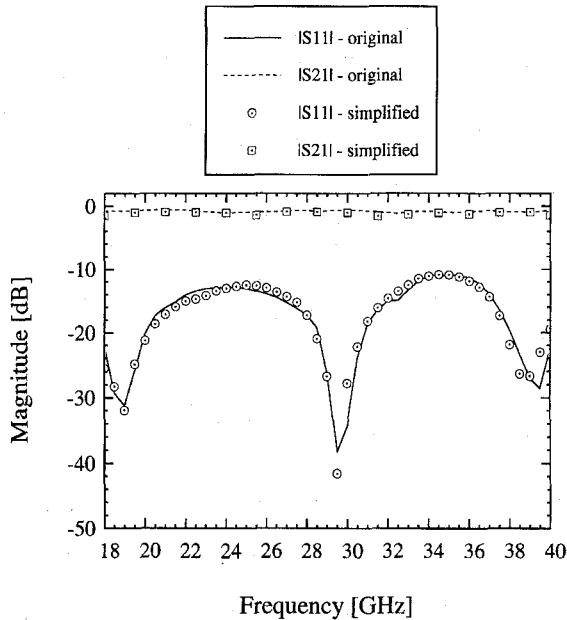


Fig. 6. Comparison between the computed S -parameters for the isolated asymmetric package with (original) and without (simplified) dc bias lines and bond wires.

except a few minor modifications for symmetric arrangement. The input/output microstrip lines and the through line are moved to the center of the package, and the two asymmetric metal filled vias are aligned on line. As it will be shown in the later sections of the paper, the symmetric arrangement of the package impacts on the performance of the package greatly. The computed scattering parameters are shown in Fig. 3. The return loss is less than -20 dB over the entire frequency range and the insertion loss remains within -1.0 dB.

B. Modeling of Isolated Asymmetric Package

In this section, the performance of the isolated asymmetric package is calculated and the computed S -parameters are shown in Fig. 4. As one can observe in the figure, the package reveals relatively good performance with better than -10.0 dB return loss over the entire frequency range. It is also noted that no cavity or spurious resonances are observed even though the size of the overall package becomes larger than half guided wavelength at frequencies above 15 GHz. The lack of any resonance is also evident from Fig. 5, which shows the computed vertical electric field distribution in the package, and is attributed to the imperfect EM shielding. Even though the package provides excellent mechanical/environmental protection and hermeticity, the side walls formed by the 12 vertical metal filled vias between the dc bias lines do not provide a solid EM shield. Leakage from the cavity semi-open walls drains the energy from the cavity and suppresses the occurrence of cavity resonances.

The effects of the bond wires and dc bias lines on the characteristics of the package are also studied. This is accomplished by computing the scattering parameters after eliminating the four bond wires between input/output RF microstrip lines and the $50\ \Omega$ through line remaining $40\ \mu\text{m}$ gap between them, and 5 dc bias lines from both sides of the package. The computed

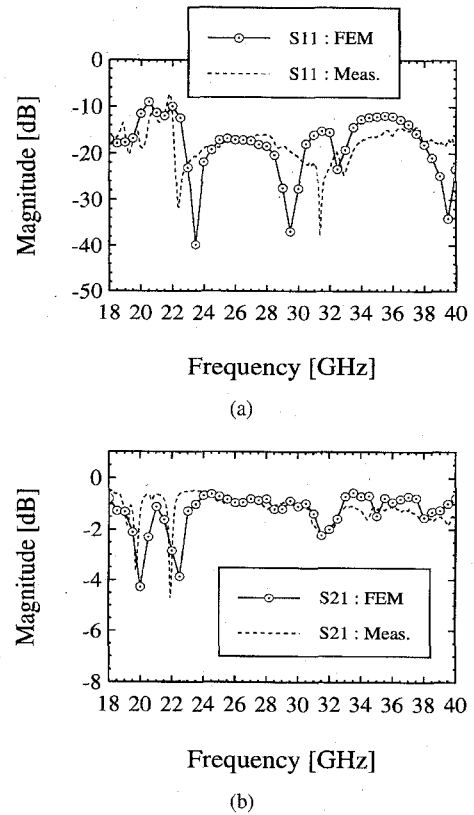


Fig. 7. Measured and computed S -parameters [(a) $|S_{11}|$ and (b) $|S_{21}|$] for the asymmetric package residing in the test fixture.

scattering parameters shown in Fig. 6 reveal no discernible differences from the previous data. This result implies that the presence of the bond wires and the dc bias lines is not critical to the performance of the package. Even further, we can argue that the shape of those structures should not disturb the overall characteristics of the package. Compared to the performance of the symmetric package shown in Fig. 3, the return loss of the asymmetric package surprisingly increases about 10 dB.

C. Modeling of Asymmetric Package Placed in a Test Fixture

The asymmetric package is now modeled by taking into consideration the test fixture to quantify the susceptibility of the package to the environment in addition to identification of the leakage and spurious resonances. The test fixture is modeled by using PEC walls along the ends of the package with an opening at the center for the input/output coaxial connectors and half height PEC walls and artificial absorbers for two other sides (see Fig. 2). The half height PEC walls are designed for the simulation of the recess depth of the test fixture and absorber is placed on top of the half PEC wall. The measured and computed scattering parameters for the package placed in the test fixture show very good agreement as illustrated in Fig. 7. As it can be observed, the overall structure including the package and test fixture suffers from spurious resonances in the low frequency region (18–24 GHz) but it exhibits very good performance in the rest of the frequency range.

The resonance phenomenon can be understood by investigating the EM field distribution at various frequencies as shown in Fig. 8. Fig. 8(a) shows the field distribution at

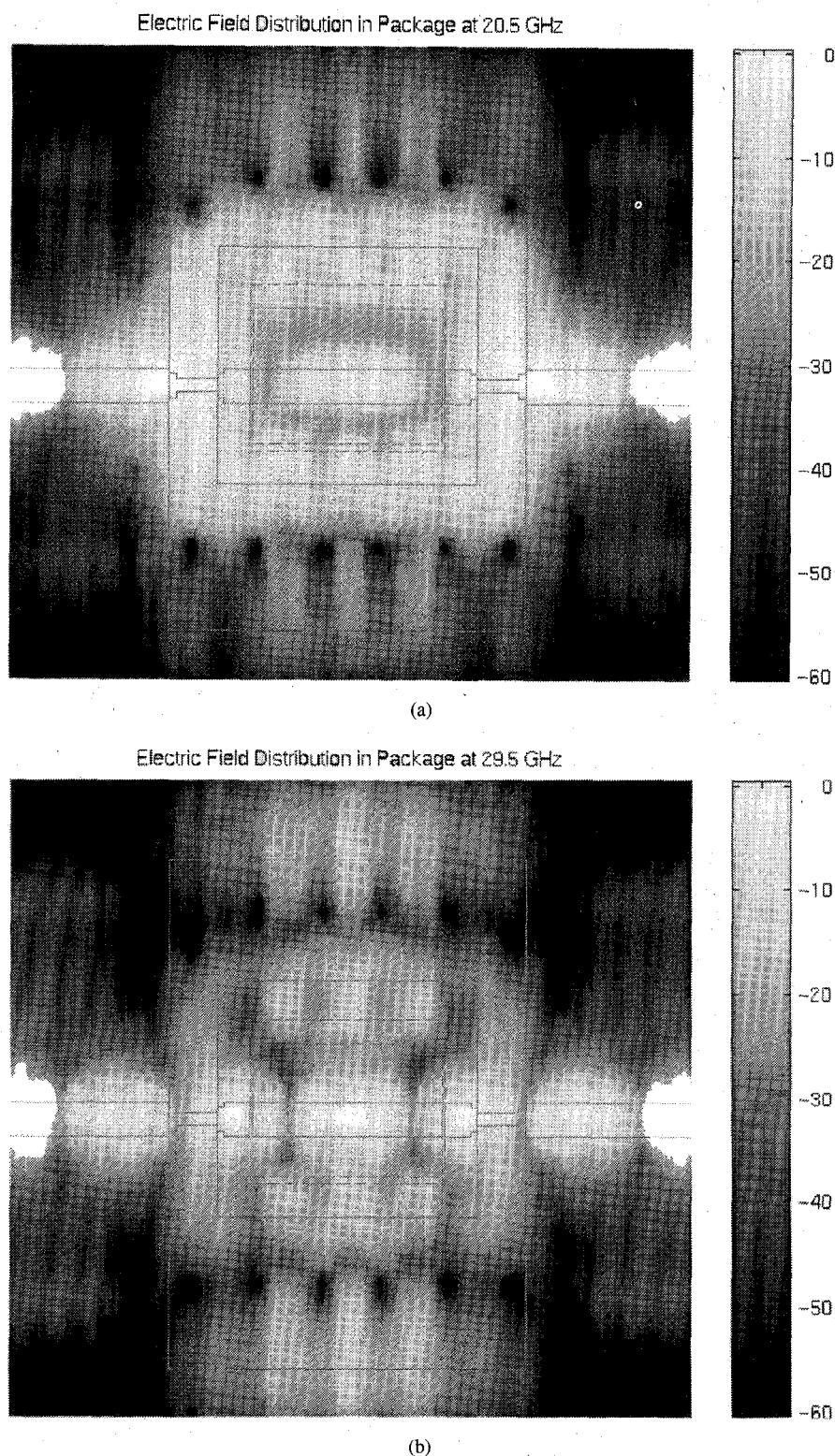


Fig. 8. Computed vertical electric field distribution (dB scale) in the asymmetric hermetic package placed in the test fixture at (a) $f = 20.5$ GHz and (b) $f = 29.5$ GHz in the isolated asymmetric hermetic package.

$f = 20.5$ GHz. At this frequency $|S_{11}|$ has a peak indicating mismatch [see Fig. 7(a)] and reveals energy leakage through the input/output RF microstrip lines and between the metal filled vias resulting in high insertion loss. It is also observed that at some frequencies the excited EM fields are strongly concentrated in the package frame which has the shape of

a dielectric ring and is placed on top of the input/output microstrip lines, along the dc bias lines. However, it should be noted that the poor EM hermeticity may cause a serious EM compatibility (EMC) problem. Fig. 8(b) shows the field distribution at 29.5 GHz, where $|S_{11}|$ has a dip indicating small insertion loss [see Fig. 7(a)]. The calculated results also

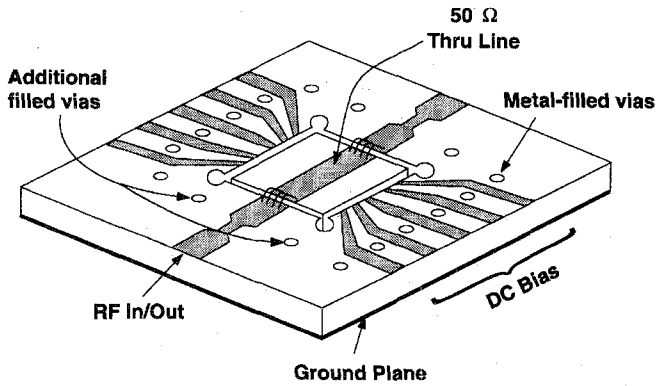


Fig. 9. Schematic diagram of the asymmetric hermetic package showing four additional vias at the input and output ends of the package.

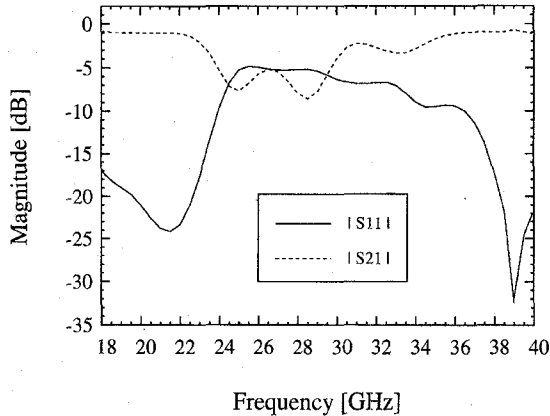


Fig. 10. Computed S -parameters for the asymmetric package with four additional vias at the input and output ends of the package.

show the voltage standing wave pattern of the EM field in the microstrip line.

IV. FEATURES WHICH CAN ENHANCE PACKAGE PERFORMANCE

A. Asymmetric Package with Additional Metal Filled Vias

To investigate the role of the vertical metal filled vias on the EM hermeticity of the package and to study the energy leakage from the input/output ends of the package, four additional vias are placed as shown in Fig. 9. The dimensions of the additional vias are the same with those of the previous set of vias. As shown in Fig. 10, the frequency response of the package is quite degraded in the most of the frequency region. Hence, it is revealed that providing additional vias could intensify package internal resonances by suppressing the leakage of the EM fields. Interestingly enough, the enforcement of strong EM hermeticity by placing additional vias causes strong unwanted internal resonances, while poor EM hermeticity makes the package more susceptible to the surrounding environment. To overcome the above difficulties, we utilized lossy packaging materials as examined in the following sections.

B. Asymmetric Package with Resistive Coating on the Inside Surface of the Lid

As an effort to suppress the unwanted resonances in the 18–24 GHz region [refer to Fig. 7(b)], a resistive sheet backed

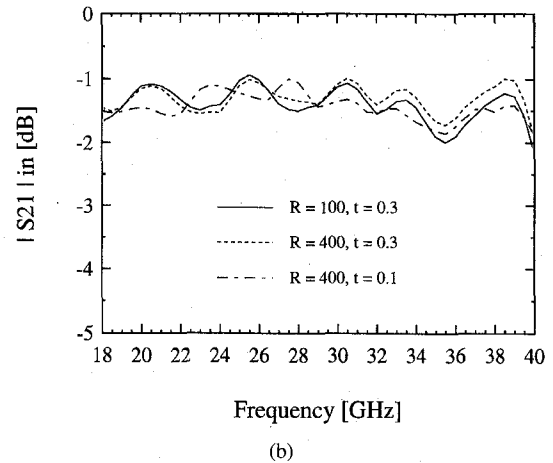
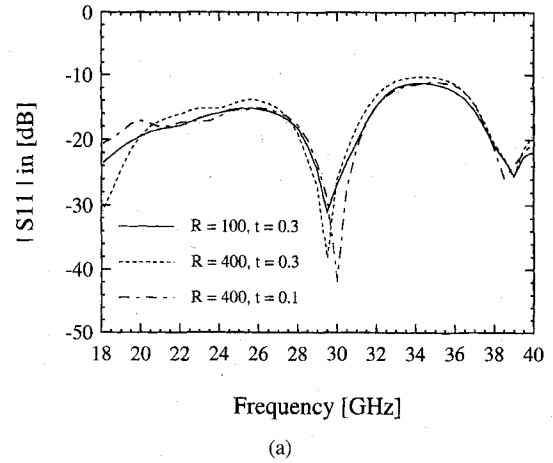


Fig. 11. Computed S -parameters [(a) $|S_{11}|$ and (b) $|S_{21}|$] for the asymmetric package residing in the test fixture with resistive coating on the inside surface of the lid. The resistivity and the thickness of the coating is indicated as R in $[\Omega/\square]$ and t in [mm], respectively.

by PEC plane is placed on top of the package. The value of the resistivity (R) of the sheet is chosen to be $100 \Omega/\square$ or $400 \Omega/\square$ and the thickness (t) to be 0.1 mm or 0.3 mm. Fig. 11 shows the computed scattering parameters. The spurious resonances in the low frequency region (18–24 GHz) are completely suppressed for all three cases and there are no distinguishable differences between them. This observation indicates that placing resistive material on the inside surface of the package lid can suppress the unwanted resonances by imposing lossy boundary condition on the fields resonating inside the package.

C. Asymmetric Package with Lossy Dielectric Frame

It is very important to suppress the internal resonances during design in order to avoid performance degradation. Suppression of the internal resonances may be accomplished through a variety of approaches. One approach may suggest the design of a leaky package so that strong resonances are not supported. This approach has been examined in the previous sections and reveals that the package interferes strongly with its surroundings. As an alternate approach, one can fabricate the package frame from a dielectric material having nonzero loss tangent. Fig. 12 shows S -parameters of the package with two different dielectric materials: Case 1: $\epsilon_r = 9.5(1.0 + j0.1)$,

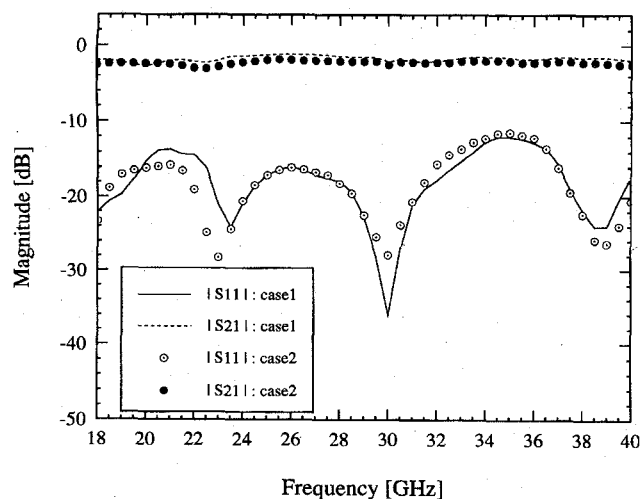


Fig. 12. Computed S -parameters for the asymmetric package with frame constructed from dielectric material with high loss tangent. Case 1: $\epsilon_r = 9.5(1.0 + j0.1)$ and $\mu_r = 1.0 + j0.0$. Case 2: $\epsilon_r = 9.5(1.0 + j0.5)$ and $\mu_r = 1.0 + j0.0$.

$\mu_r = 1.0 + j0.0$, and Case 2: $\epsilon_r = 9.5(1.0 + j0.5)$, $\mu_r = 1.0 + j0.0$. As one can realize from the figure, the unwanted cavity resonances in the low frequency region are completely suppressed by the assigned small amount of loss in the packaging material. It is also important to notice that the insertion loss increases to -2 dB due to the loss in the packaging material. The differences in the scattering parameters for two cases corresponding to different loss tangents remain negligible in the whole frequency spectrum.

V. CONCLUSION

In this paper, a 18–40 GHz hermetic package is experimentally characterized and also modeled using finite element method. The FEM accurately predicts the S -parameters, energy leakage, and spurious resonances which degrade the package performance. To the best of our knowledge this is the first comprehensive study of RF leakage and resonances in a millimeter-wave package. The modeled results show that the unwanted spurious resonances in the package can be suppressed by incorporating a resistive coating on the lid and by the use of dielectric materials with high loss tangent. The modeled S -parameters for a symmetric package predict low insertion loss on the order of -1.0 dB over the 18–40 GHz band. From this study, it is clear that symmetry in the package construction is crucial for good performance.

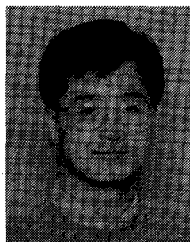
ACKNOWLEDGMENT

The authors would like to thank the Maui High Performance Computing Center (MHPCC), the Army Research Office (ARO), the Army Research Laboratory (ARL), and the University of Michigan Center for Parallel Computing (CPC).

REFERENCES

- [1] L. P. B. Katehi, "The role of EM modeling in integrated packaging," in *1993 IEEE AP-S Dig.*, July 1993, pp. 982–985.
- [2] H. J. Kuno and T. A. Midford, "The evolution of MMIC packaging," in *1993 IEEE AP-S Dig.*, July 1993, pp. 1005–1008.
- [3] D. W. Griffin and A. J. Parfitt, "Electromagnetic design aspects of packages for phased array modules that may incorporate monolithic antenna elements," in *1993 IEEE AP-S Dig.*, July 1993, pp. 986–989.

- [4] B. A. Ziegner, "High performance MMIC hermetic packaging," *Microwave J.*, pp. 133–139, Nov. 1986.
- [5] H. Bierman, "Designers strive for low cost MMIC packages," *Microwave J.*, pp. 100–106, Sept. 1992.
- [6] T. Shibata, S. Kimura, H. Kimura, Y. Imai, Y. Umeda, and Y. Akazawa, "Design technique for a 60 GHz-bandwidth distributed baseband amplifier IC module," *IEEE J. Solid-State Circ.*, vol. 29, pp. 1537–1544, Dec. 1994.
- [7] J.-G. Yook, N. Dib, E. Yasan, and L. Katehi, "A study of hermetic transitions for microwave packages," in *1995 IEEE MTT-S Int. Microwave Symp. Dig.*, May 1995, pp. 1579–1582.
- [8] J. Gippich, L. Dickens, B. Hayes, and F. Sacks, "A Compact 8–14 GHz LTCC Stripline coupler network for high efficiency power combining with better than 82% combining efficiency," in *1995 IEEE MTT-S Int. Microwave Symp. Dig.*, May 1995, pp. 1583–1586.
- [9] M. Rittweger, M. Werthen, J. Kunisch, I. Wolff, P. Chall, B. Balm, and P. Lok, "3D FDTD Analysis of a SOT353 package containing a bipolar wideband cascode transistor using compression approach," in *1995 IEEE MTT-S Int. Microwave Symp. Dig.*, May 1995, pp. 1587–1590.
- [10] J.-G. Yook, N. Dib, and L. Katehi, "Characterization of high frequency interconnects using finite difference time domain and finite element methods," *IEEE Trans. Microwave Theory Tech.*, pp. 1727–1736, vol. 42, Sept. 1994.
- [11] J.-G. Yook and L. Katehi, "Characterization of MIMIC's packages using a parallelized 3D FEM code," *1996 ACES Symp. Proc.*, Monterey, CA.



Jong-Gwan Yook (S'86) was born in Korea in 1964. He received the B.S. and M.S. degrees in electronic engineering from Yonsei University, Seoul, Korea, in 1987 and 1989, respectively. He is currently working toward the Ph.D. degree in electrical engineering at the University of Michigan, Ann Arbor.

He is also working as a Research Assistant in the University of Michigan's Radiation Laboratory. His main research interests are in the area of electromagnetic characterization of VLSI and MMIC

interconnects using the finite element method and development of numerical techniques for analysis and design of high frequency circuits with emphasis on parallel computing.



Linda P. B. Katehi (S'81–M'84–SM'89–F'95) received the B.S.E.E. degree from the National Technical University of Athens, Greece, in 1977, and the M.S.E.E. and Ph.D. degrees from the University of California at Los Angeles in 1981, and 1984, respectively.

Since 1984 she has been a faculty member at the University of Michigan at Ann Arbor. Her research interests have focused on the development and characterization (theoretical and experimental) of microwave millimeter printed circuits, the computer-

aided design of VLSI interconnects, the development and characterization of micromachined circuits for millimeter-wave and submillimeter-wave applications and the development of low-loss lines for Terahertz-frequency applications. She has also been researching theoretically and experimentally various types of uniplanar radiating structures for hybrid-monolithic and monolithic oscillator and mixer designs. She has been the author and co-author of more than 220 papers published in refereed journals and symposia proceedings. She is an Associate Editor of the IEEE TRANSACTIONS ON MICROWAVE THEORY AND TECHNIQUES and the IEEE TRANSACTIONS ON ANTENNAS AND PROPAGATION.

Dr. Katehi received the IEEE AP-S W. P. King (Best Paper Award for a Young Engineer) in 1984, the IEEE AP-S S. A. Schellkunoff Award (Best Paper Award) in 1985, the NSF Presidential Young Investigator Award and an URSI Young Scientist Fellowship in 1987, the Humboldt Research Award and the University of Michigan Faculty Recognition Award in 1994, and the IEEE MTT-S Microwave Prize in 1996. She is a member of the IEEE AP-S, MTT-S, Sigma Xi, Hybrid Microelectronics, URSI Commission D, and the AP-S ADCOM (1992–1995).



Rainee N. Simons (S'76-M'80-SM'89) received the B.S. degree in electronics and communications engineering from Mysore University, India, and the M.Tech. degree in electronics and communications engineering from the Indian Institute of Technology, Kharagpur, in 1972 and 1974, respectively. In 1983 he received the Ph.D. degree in electrical engineering from the Indian Institute of Technology (IIT), New Delhi.

In 1979 he joined IIT as a Senior Scientific Officer, where he worked on fineline components for millimeter-wave applications and also on the X-band toroidal latching ferrite phase shifters for phased arrays. Since 1985 he has been with the Space Electronics/Communications Division of NASA Lewis Research Center, Cleveland, OH, as a National Research Council Research Associate (1985-1987), a CWRU Research Associate (1987-1990), a Senior Engineer with Sverdrup Technology, Inc. (1990-1993), and a Senior Engineer with Nydma, Inc. (1994-present). His work at NASA has focused on optical control of MESFET and HEMT devices, high-temperature superconductivity, modeling of co-planar waveguide discontinuities, CPW feed systems for printed antennas, linearly tapered slot antenna arrays for communications and packaging of MMIC's. He is the author of *Optical Control of Microwave Devices* (Artech House) and "High-Temperature Superconducting Coplanar Waveguide Microwave Circuits and Antennas," a chapter in *Advances in High-Tc Superconductors* (Switzerland: Trans-Tech Publications). He is an Editorial Board member of IEEE TRANSACTIONS ON MICROWAVE THEORY AND TECHNIQUES.

Dr. Simons has received the distinguished alumni award from his alma mater and several NASA Certificate of Recognition and Group Achievement awards. He organized a Special Session on "Advances in MMIC Packaging for Phased Array Antennas" at the 1993 IEEE AP-S International Symposium and URSI Radio Science Meeting. He served as the Guest Editor for the Special Issue on "Packaging Technologies for Phased Array Applications" in of the September 1995 IEEE TRANSACTIONS ON ANTENNAS AND PROPAGATION.

Kurt A. Shalkhauser, photograph and biography not available at the time of publication.